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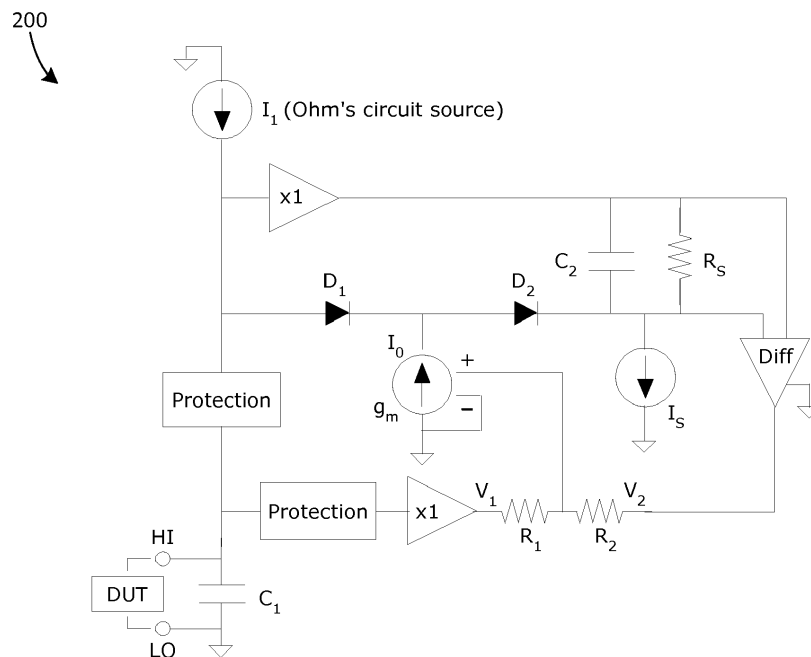
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(54) **VOLTAGE CLAMP**

(57) A voltage clamp circuit (200) which operates using a voltage controlled current source (I<sub>0</sub>) where the change of the polarity of the voltage controlled current source controls whether it is clamping or not. While clamping, the stability of the control loop uses the capacitance of the output (C<sub>1</sub>) to create a single pole roll-off

of the loop gain and while not clamping, uses the capacitance (C<sub>2</sub>) of the circuit which sets the clamping voltage to produce the roll-off. The circuit operates in a linear fashion both while clamping and not clamping, which allows for a faster response when clamping is needed.



**FIG. 2**

## Description

### Technical Field

[0001] This disclosure relates generally to voltage clamps and, more particularly, to voltage clamps designed for dry circuit ohms clamping.

### Brief Description of the Drawings

#### [0002]

FIGURE 1 illustrates an example of a loop concept circuit on which certain voltage clamp design embodiments in accordance with the disclosed technology are based.

FIGURE 2 is a block diagram illustrating an example of a voltage clamp circuit coupled with a device under test (DUT) in accordance with certain embodiments of the disclosed technology.

FIGURE 3 is a block diagram illustrating an alternative example of a voltage clamp circuit coupled with a DUT in accordance with certain embodiments of the disclosed technology.

### Detailed Description

[0003] Dry circuit resistance generally requires a sourcing circuit that limits the maximum applied voltage across the device under test (DUT) to less than a few tens of millivolts (e.g., 20mV to 30mV). This is typically needed to prevent the breaking down of a thin oxide that can form on electrical contacts within the DUT, such as a relay contact or a contact within a connector pair.

[0004] Embodiments of the disclosed technology are generally directed to a circuit having a maximum voltage limit and a maximum current limit. The circuit may include a first current source configured to drive the output of the circuit (i.e., by setting the maximum current) and a second current source having a polarity that is opposite the polarity of the first current source, and is coupled with a resistance.

[0005] A voltage-controlled current source may be configured to, when its current has a polarity opposite that of the first current source, remove current from the output and, when its current has a polarity that is opposite that of the second current source, remove current from the node common to the second current source and the resistance, thus reducing the current through the resistance.

[0006] The circuit may include a component configured to compare the output voltage to the voltage across the resistance and drive the voltage controlled current source's input such that there is a negative feedback.

[0007] The circuit may include a first capacitance added to the output of the circuit to provide a single pole roll off of the output voltage from the input to the voltage controlled current source. The circuit may also include a

second capacitance added across the resistance to provide a single pole roll off of the voltage across the resistance from the input to the voltage controlled current source.

5 [0008] FIGURE 1 illustrates an example of a loop concept circuit 100 on which certain voltage clamp design embodiments in accordance with the disclosed technology are based. In the example 100, a capacitor C causes  
10 a 20dB/dec roll-off where  $\frac{g_m}{C}$  represents the gain-bandwidth of the circuit. The current source  $g_m$  should have a constant gain for frequencies less than or equal to  $\frac{g_m}{C}$ .

15 [0009] FIGURE 2 is a block diagram illustrating an example of a voltage clamp circuit 200 coupled with a DUT in accordance with certain embodiments of the disclosed technology. In the example, a current source  $I_0$  is steered by the two diodes  $D_1$  and  $D_2$  to regulate the voltage on one of two capacitors  $C_1$  or  $C_2$ . When the  $I_0$  current is negative, the Ohm's current source  $I_1$  is diverted through the first diode  $D_1$ , thus reducing the current flowing to the HI connection.

20 [0010] When the  $I_0$  current is positive, however, the  $I_S$  current source is supplied through the second diode  $D_2$ , thus reducing the current flowing through a source resistor  $R_S$ . The loop circuit 200 generally either regulates

25  $\frac{V_1}{R_1}$  to be equal to  $\frac{-V_2}{R_2}$  (i.e., with the first diode  $D_1$  conducting) or regulates  $\frac{V_2}{R_2}$  to be equal to  $\frac{-V_1}{R_1}$  (i.e., with the second diode  $D_2$  conducting).

30 [0011] In the example 200, a voltage  $V_1$  may be clamped to  $\frac{R_1}{R_2} R_S I_S$ . Thus,  $\frac{R_1}{R_2} R_S I_S$  generally needs to be set between 20mV and 30mV to meet the  
35 needs for dry circuit testing. The loop bandwidth is  $\frac{g_m}{C_1}$

40 when  $D_1$  is conducting and switches to  $\frac{g_m}{C_2}$  when  $D_2$  is conducting.

45 [0012] The ohm's current source protection will generally drop voltage when current is flowing through it. Therefore, the capacitor  $C_2$  and source resistor  $R_S$  may be bootstrapped to follow the ohm's current source's voltage. This advantageously prevents both of the diodes  $D_1$  and  $D_2$  from turning on at the same time. A differential stage, Diff, may be used to translate the voltage across the source resistor  $R_S$  to ground to be mixed against the voltage  $V_1$ .

50 [0013] In the example 200, the first capacitor  $C_1$  is used to limit the rate that the voltage  $V_1$  rises when a conduction path (e.g., the DUT) between the HI and LO connections is suddenly removed, e.g., to minimize the over-

shoot. Using the capacitance- $g_m$  interaction to control stability generally avoids problems that the capacitance may present in situations where a normal voltage source loop is used.

**[0014]** FIGURE 3 is a block diagram illustrating an alternative example of a voltage clamp circuit 300 coupled with a DUT in accordance with certain embodiments of the disclosed technology. In the example 300, the clamp voltage for  $V_1$  is  $I_S * R_S$ .

**[0015]** Examples provide a voltage clamp circuit, comprising a first current source configured to drive an output of the circuit, the first current source having a polarity, a second current source coupled with a resistance, the second current source having a polarity that is opposite the polarity of the first current source, a voltage-controlled current source having a polarity and configured to, when its polarity is opposite that of the first current source, remove current from the output and, when its polarity is opposite that of the second current source, remove current from a node common to the second current source and the resistance, a first capacitance coupled with the output, and a second capacitance coupled across the resistance.

**[0016]** Some examples further comprise a component configured to compare the output voltage to the voltage across the resistance.

**[0017]** In some examples the component is further configured to drive an input of the voltage-controlled current source such that there is a negative feedback.

**[0018]** In some examples the first current source is configured to drive the output of the circuit by setting a maximum current.

**[0019]** In some examples the first capacitance is configured to provide a single pole roll off of the output voltage from the input to the voltage-controlled current source.

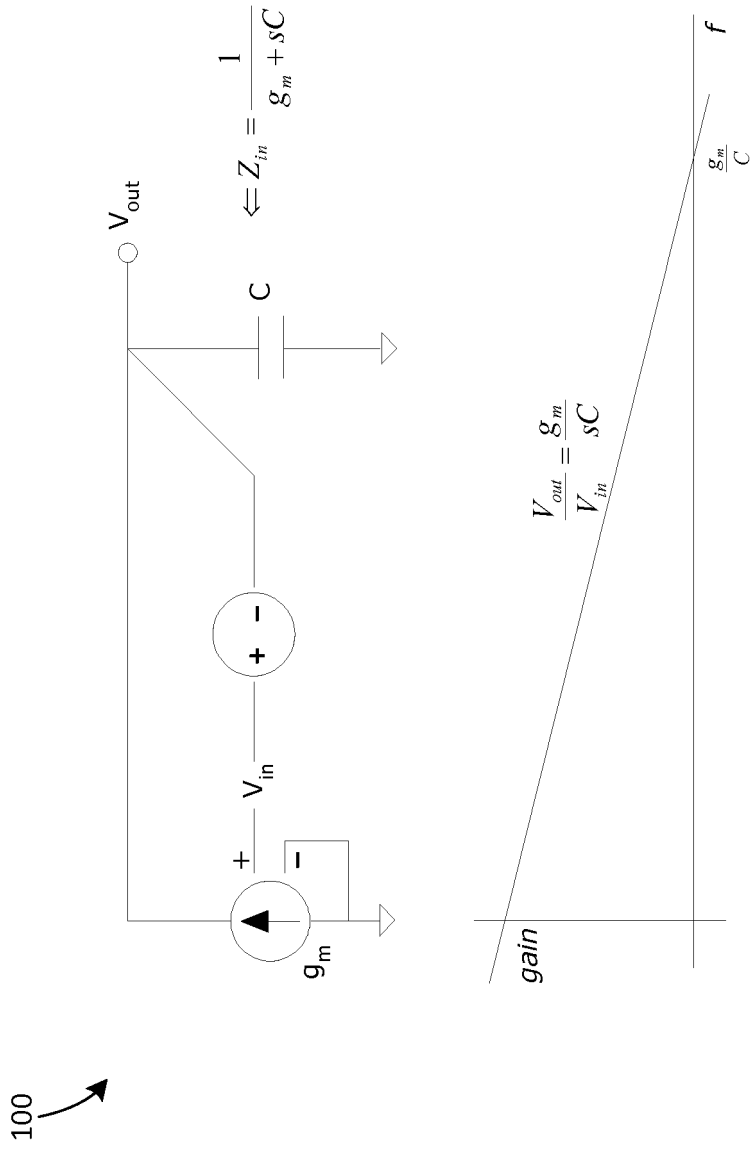
**[0020]** In some examples the second capacitance is configured to provide a single pole roll off of the voltage across the resistance from input to the voltage controlled current source.

**[0021]** Having described and illustrated the principles of the invention with reference to illustrated embodiments, it will be recognized that the illustrated embodiments may be modified in arrangement and detail without departing from such principles, and may be combined in any desired manner. And although the foregoing discussion has focused on particular embodiments, other configurations are contemplated.

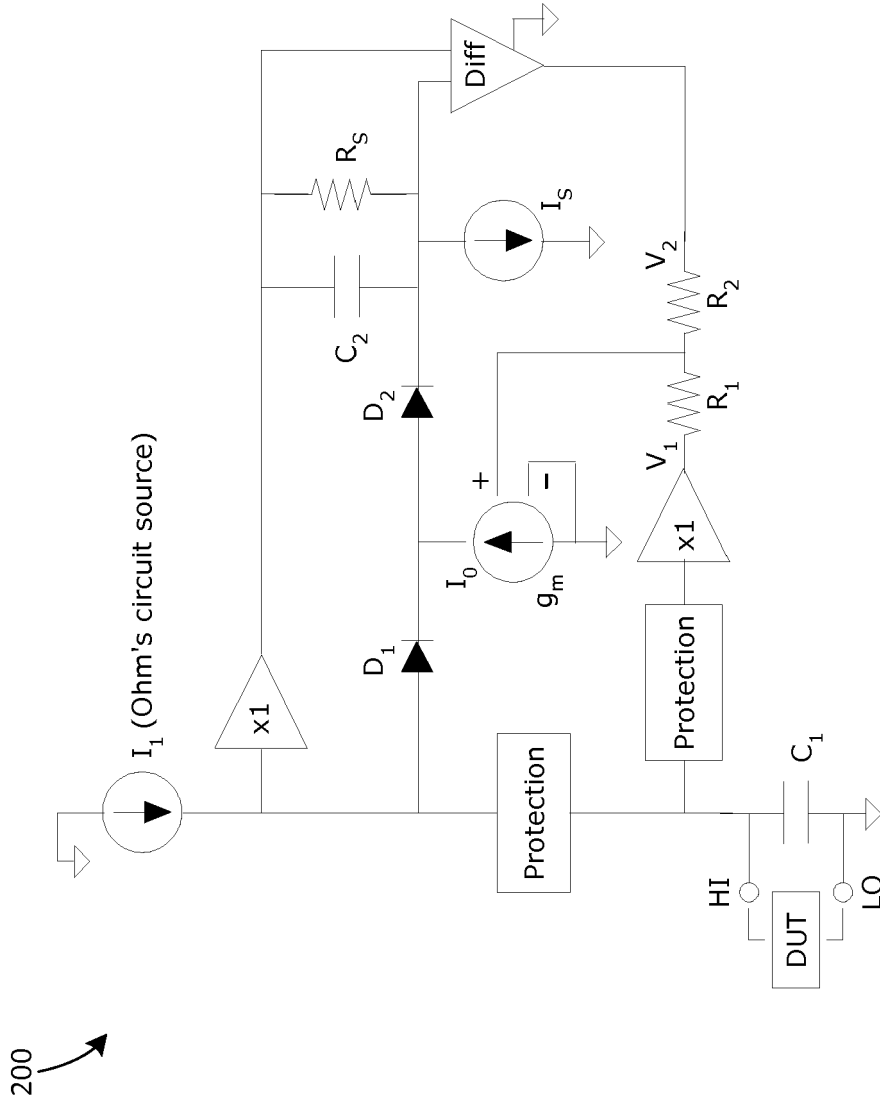
**[0022]** In particular, even though expressions such as "according to an embodiment of the invention" or the like are used herein, these phrases are meant to generally reference embodiment possibilities, and are not intended to limit the invention to particular embodiment configurations. As used herein, these terms may reference the same or different embodiments that are combinable into other embodiments.

## Claims

1. A voltage clamp circuit, comprising:
  - a first current source configured to drive an output of the circuit, the first current source having a polarity;
  - a second current source coupled with a resistance, the second current source having a polarity that is opposite the polarity of the first current source;
  - a voltage-controlled current source having a polarity and configured to:
    - when its polarity is opposite that of the first current source, remove current from the output; and
    - when its polarity is opposite that of the second current source, remove current from a node common to the second current source and the resistance;
  - a first capacitance coupled with the output; and
  - a second capacitance coupled across the resistance.
2. The circuit of claim 1, further comprising a component configured to compare the output voltage to the voltage across the resistance.
3. The circuit of claim 2, wherein the component is further configured to drive an input of the voltage-controlled current source such that there is a negative feedback.
4. The circuit of any previous claim, wherein the first current source is configured to drive the output of the circuit by setting a maximum current.
5. The circuit of any previous claim, wherein the first capacitance is configured to provide a single pole roll off of the output voltage from the input to the voltage-controlled current source.
6. The circuit of any of claims 1 to 4, wherein the second capacitance is configured to provide a single pole roll off of the voltage across the resistance from input to the voltage controlled current source.



**FIG. 1**



**FIG. 2**

300

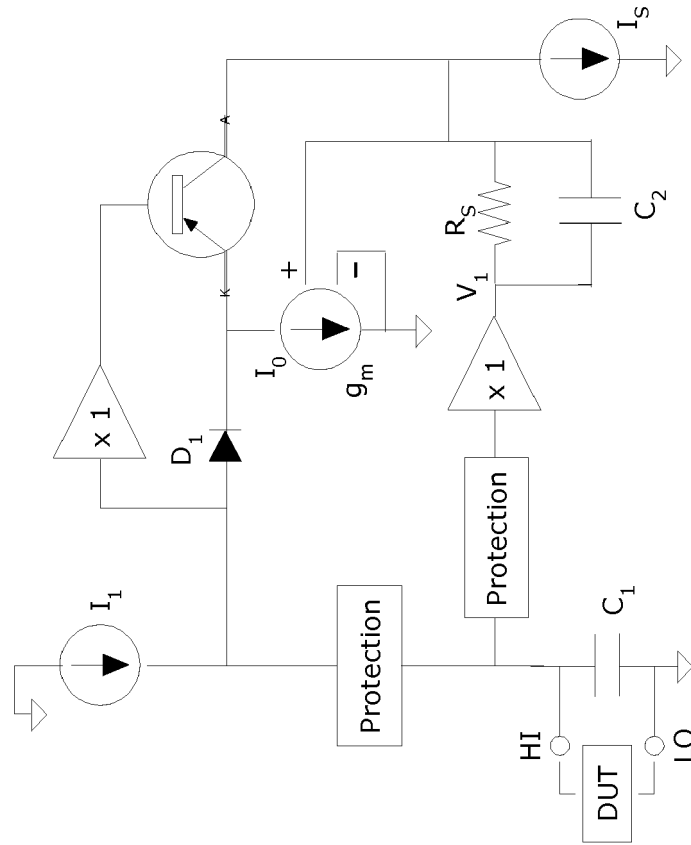


FIG. 3



EUROPEAN SEARCH REPORT

Application Number  
EP 15 19 5715

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DOCUMENTS CONSIDERED TO BE RELEVANT			
Category	Citation of document with indication, where appropriate, of relevant passages	Relevant to claim	CLASSIFICATION OF THE APPLICATION (IPC)
X	US 5 337 205 A (HAUN ANDY A [US] ET AL) 9 August 1994 (1994-08-09) * column 4, line 48 - line 55 * * figure 1 * * figure 2 * * column 3, line 53 - line 55 * * figure 3 * * column 4, line 10 - line 13 * * column 4, line 56 - line 65 * -----	1-6	INV. G01R1/36  ADD. H02H9/04
A	US 6 717 450 B1 (LINDER LLOYD F [US]) 6 April 2004 (2004-04-06) * figure 1 * -----	1	
A	US 5 384 532 A (UHLING THOMAS F [US]) 24 January 1995 (1995-01-24) * column 9, line 55 - line 58 * -----	5,6	
A	US 2007/216391 A1 (BLANKEN PIETER G [NL]) 20 September 2007 (2007-09-20) * the whole document * -----	4	TECHNICAL FIELDS SEARCHED (IPC)
A	EP 1 832 888 A1 (TERADYNE INC [US]) 12 September 2007 (2007-09-12) * the whole document * -----	1	G01R H01L H02H
A	US 2009/086395 A1 (SKRENES LAWRENCE [US] ET AL) 2 April 2009 (2009-04-02) * the whole document * -----	1	
The present search report has been drawn up for all claims			
Place of search Munich		Date of completion of the search 15 April 2016	Examiner Meliani, Chafik
CATEGORY OF CITED DOCUMENTS X : particularly relevant if taken alone Y : particularly relevant if combined with another document of the same category A : technological background O : non-written disclosure P : intermediate document		T : theory or principle underlying the invention E : earlier patent document, but published on, or after the filing date D : document cited in the application L : document cited for other reasons ..... & : member of the same patent family, corresponding document	

EPO FORM 1503 03.82 (P04C01)

ANNEX TO THE EUROPEAN SEARCH REPORT  
ON EUROPEAN PATENT APPLICATION NO.

EP 15 19 5715

5 This annex lists the patent family members relating to the patent documents cited in the above-mentioned European search report.  
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15-04-2016

Patent document cited in search report	Publication date	Patent family member(s)	Publication date
US 5337205 A	09-08-1994	CA 2122259 A1	17-03-1994
		DE 69308534 D1	10-04-1997
		DE 69308534 T2	11-09-1997
		EP 0611495 A1	24-08-1994
		JP 2801404 B2	21-09-1998
		JP H07501200 A	02-02-1995
		US 5337205 A	09-08-1994
		WO 9406190 A1	17-03-1994
US 6717450 B1	06-04-2004	AU 2003269419 A1	02-12-2003
		TW 200402541 A	16-02-2004
		US 6717450 B1	06-04-2004
		WO 03098231 A2	27-11-2003
US 5384532 A	24-01-1995	NONE	
US 2007216391 A1	20-09-2007	AT 544225 T	15-02-2012
		CN 101027621 A	29-08-2007
		EP 1797492 A1	20-06-2007
		ES 2381482 T3	28-05-2012
		JP 4734511 B2	27-07-2011
		JP 2008515077 A	08-05-2008
		KR 20070059201 A	11-06-2007
		US 2007216391 A1	20-09-2007
		WO 2006035394 A1	06-04-2006
EP 1832888 A1	12-09-2007	AT 434189 T	15-07-2009
		EP 1832888 A1	12-09-2007
		TW 200817706 A	16-04-2008
		US 2007210810 A1	13-09-2007
		WO 2007102856 A1	13-09-2007
US 2009086395 A1	02-04-2009	US 2009086395 A1	02-04-2009
		WO 2009046013 A1	09-04-2009